Source Separation Tutorial Mini-Series II: Introduction to Non-Negative Matrix Factorization

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Center for Computer Research in Music and Acoustics, Stanford University

> DSP Seminar April 9th, 2013

Roadmap of Talk

1 Motivation

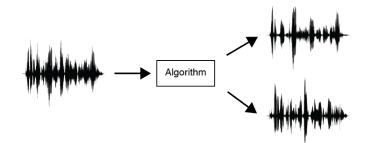
- **2** Current Approaches
- 3 Non-Negative Matrix Factorization (NMF)
- **4** Source Separation via NMF
- **5** Algorithms for NMF
- 6 Matlab Code

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General Idea



Music Remixing and Content Creation

Music remixing and content creation



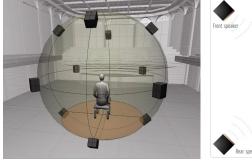
Audio Post-Production and Remastering

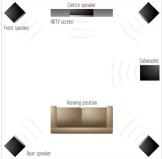
Audio post-production and remastering



Spatial Audio and Upmixing

Spatial audio and upmixing





Denoising

Denoising

- Separate noise speech
- Remove background music from music
- Remove bleed from other instruments



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Current Approaches I

• Microphone Arrays

- Beamforming to "listen" in a particular direction [BCH08]
- Requires multiple microphones
- Adaptive Signal Processing
 - Self-adjusting filter to remove an unwanted signal [WS85]
 - Requires knowing the interfering signal
- Independent Component Analysis
 - Leverages statistical independence between signals [HO00]
 - Requires N recordings to separate N sources

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Current Approaches II

- Computational Auditory Scene Analysis
 - Leverages knowledge of auditory system [WB06]
 - Still requires some other underlying algorithm
- Sinusoidal Modeling
 - Decomposes sound into sinusoidal peak tracks [Smi11, Wan94]
 - Problem in assigning sound source to peak tracks
- Classical Denoising and Enhancement
 - Wiener filtering, spectral subtraction, MMSE STSA (Talk 1)
 - Difficulty with time varying noise

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Current Approaches III

- Non-Negative Matrix Factorization & Probabilistic Models
 - Popular technique for processing audio, image, text, etc.
 - Models spectrogram data as mixture of prototypical spectra
 - Relatively compact and easy to code algorithms
 - Amenable to machine learning
 - In many cases, works surprisingly well
 - The topic of today's discussion!

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Matrix Factorization

• Decompose a matrix as a product of two or more matrices

$\mathbf{A} = \mathbf{B} \, \mathbf{C} \qquad \qquad \mathbf{A} \approx \mathbf{B} \, \mathbf{C}$ $\mathbf{D} = \mathbf{E} \, \mathbf{F} \, \mathbf{G} \qquad \qquad \mathbf{D} \approx \mathbf{E} \, \mathbf{F} \, \mathbf{G}$

- Matrices have special properties depending on factorization
- Example factorizations:
 - Singular Value Decomposition (SVD)
 - Eigenvalue Decomposition
 - QR Decomposition (QR)
 - Lower Upper Decomposition (LU)
 - Non-Negative Matrix Factorization

Matrix Factorization

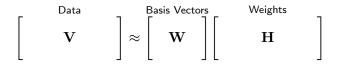
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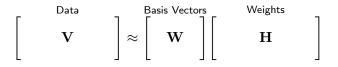
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Non-Negative Matrix Factorization



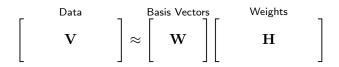
- A matrix factorization where everything is non-negative
- $\mathbf{V} \in \mathbf{R}^{F imes T}_+$ original non-negative data
- $\mathbf{W} \in \mathbf{R}_+^{F imes K}$ matrix of basis vectors, dictionary elements
- $\mathbf{H} \in \mathbf{R}_+^{K imes T}$ matrix of activations, weights, or gains
- K < F < T (typically)
 - A compressed representation of the data
 - A low-rank approximation to V

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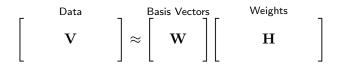
Interpretation of ${\bf V}$



•
$$\mathbf{V} \in \mathrm{R}^{F imes T}_+$$
 - original non-negative data

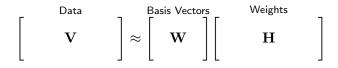
- Each column is an F-dimensional data sample
- Each row represents a data feature
- We will use audio spectrogram data as ${\bf V}$

Interpretation of ${\bf W}$



- $\mathbf{W} \in \mathrm{R}^{F imes K}_+$ matrix of basis vectors, dictionary elements
 - A single column is referred to as a basis vector
 - Not orthonormal, but commonly normalized to one

Interpretation of ${\bf H}$



- $\mathbf{H} \in \mathrm{R}^{K imes T}_+$ matrix of activations, weights, or gains
 - A row represents the gain of corresponding basis vector
 - Not orthonormal, but commonly normalized to one

NMF With Spectrogram Data

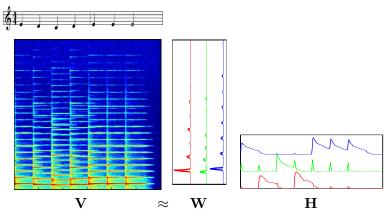


Figure : NMF of Mary Had a Little Lamb with K = 3 play stop

- The basis vectors capture prototypical spectra [SB03]
- The weights capture the gain of the basis vectors

NMF With Spectrogram Data

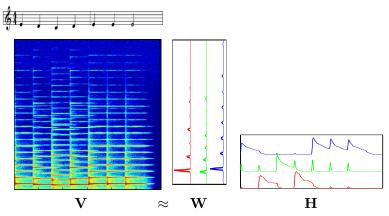
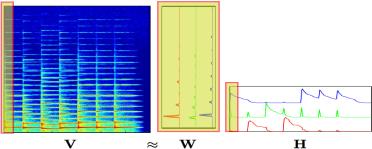


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Factorization Interpretation I

Columns of $\mathbf{V}\approx$ as a weighted sum (mixture) of basis vectors

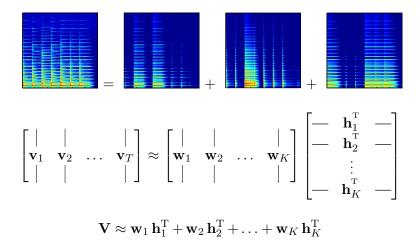


 \approx

$$\begin{bmatrix} | & | & | \\ \mathbf{v}_1 & \mathbf{v}_2 & \dots & \mathbf{v}_T \\ | & | & | \end{bmatrix} \approx \begin{bmatrix} K \\ \sum_{j=1}^K \mathbf{H}_{j1} \mathbf{w}_j & \sum_{j=1}^K \mathbf{H}_{j2} \mathbf{w}_j & \dots & \sum_{j=1}^K \mathbf{H}_{jT} \mathbf{w}_j \end{bmatrix}$$

Factorization Interpretation II

 ${\bf V}$ is approximated as sum of matrix "layers"



Questions

• How do we use ${\bf W}$ and ${\bf H}$ to perform separation?

• How do we solve for W and H, given a known V?

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• How do we use $\mathbf W$ and $\mathbf H$ to perform separation?

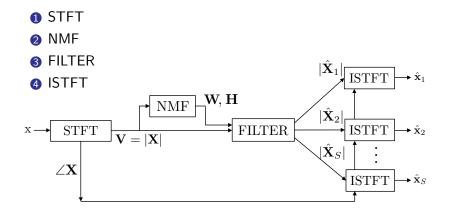
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Roadmap of Talk

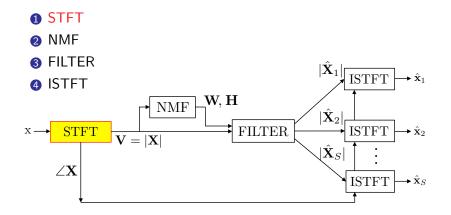
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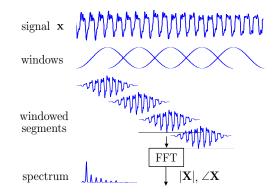
General Separation Pipeline



General Separation Pipeline



Short-Time Fourier Transform (STFT)



- Inputs time domain signal x
- Outputs magnitude $|\, {\bf X} \,|$ and phase $\angle \, {\bf X}$ matrices

Short-Time Fourier Transform (STFT)

$$X_m(\omega_k) = e^{-j\omega_k mR} \sum_{n=-N/2}^{N/2-1} x(n+mR)w(n)e^{-j\omega_k n}$$

$$x(n) =$$
 input signal at time n

$$w(n) \;\;=\;\;$$
 length M window function (e.g. Hann, etc.)

$$N = \mathsf{DFT}$$
 size, in samples

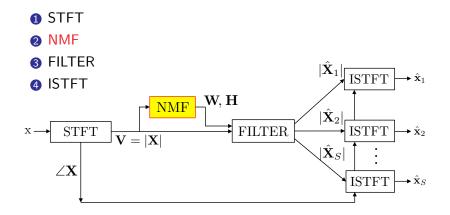
$$R =$$
 hop size, in samples, between successive DFT

$$M =$$
 window size, in samples

$$w_k = 2\pi k/N, \ k = 0, 1, 2, \dots, N-1$$

- · Choose window, window size, DFT size, and hop size
- Must maintain constant overlap-add COLA(R) [Smi11]

General Separation Pipeline

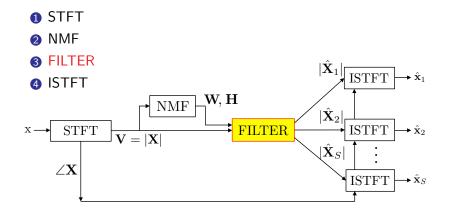


Non-Negative Matrix Factorization

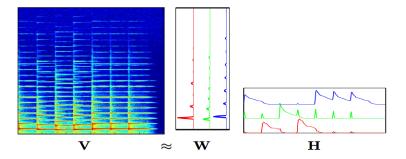
- Inputs $|\, {\bf X} \, |,$ outputs ${\bf W}$ and ${\bf H}$

• Algorithm to be discussed

General Separation Pipeline



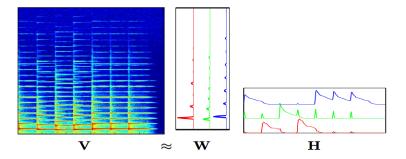
Source Synthesis I



- Choose a subset of basis vectors \mathbf{W}_s and activations \mathbf{H}_s to reconstruct source s
- Estimate the source *s* magnitude:

$$|\hat{\mathbf{X}}_{s}| = \mathbf{W}_{s} \, \mathbf{H}_{s} = \sum_{i \in s} (\mathbf{w}_{i} \, \mathbf{h}_{i}^{\mathrm{T}})$$

Source Synthesis I

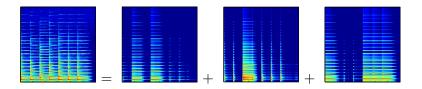


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Source Synthesis II

Example 1: "D" pitches as a single source



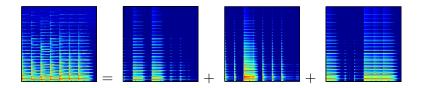
$$\mathbf{V} \approx \mathbf{w}_1 \, \mathbf{h}_1^{\mathrm{T}} + \mathbf{w}_2 \, \mathbf{h}_2^{\mathrm{T}} + \mathbf{w}_3 \, \mathbf{h}_3^{\mathrm{T}}$$

•
$$|\hat{\mathbf{X}}_s| \approx \mathbf{w}_1 \, \mathbf{h}_1^{\mathrm{T}}$$

• Use one basis vector to reconstruct a source

Source Synthesis III

Example 2: "D" and "E" pitches as a source



$$\mathbf{V} \approx \mathbf{w}_1 \, \mathbf{h}_1^{\mathrm{T}} + \mathbf{w}_2 \, \mathbf{h}_2^{\mathrm{T}} + \mathbf{w}_3 \, \mathbf{h}_3^{\mathrm{T}}$$

•
$$|\hat{\mathbf{X}}_s| \approx \mathbf{w}_1 \, \mathbf{h}_1^{\mathrm{T}} + \mathbf{w}_2 \, \mathbf{h}_2^{\mathrm{T}}$$

• Use two (or more) basis vector to reconstruct a source

Source Filtering I

Alternatively, we can estimate $|\hat{\mathbf{X}}_s|$ by filtering $|\,\mathbf{X}\,|$ via:

1 Generate a filter $\mathbf{M}_s, \ \forall s$

$$\mathbf{M}_{s} = \frac{(\mathbf{W}_{s} \mathbf{H}_{s})^{\alpha}}{\sum\limits_{i=1}^{K} (\mathbf{W}_{i} \mathbf{H}_{i})^{\alpha}} = \frac{|\hat{\mathbf{X}}_{s}|^{\alpha}}{\sum\limits_{i=1}^{K} |\hat{\mathbf{X}}_{i}|^{\alpha}} = \frac{\sum\limits_{i \in s}^{K} (\mathbf{w}_{i} \mathbf{h}_{i}^{\mathrm{T}})^{\alpha}}{\sum\limits_{i=1}^{K} (\mathbf{w}_{i} \mathbf{h}_{i}^{\mathrm{T}})^{\alpha}}$$

where $\alpha \in \mathbf{R}_+$ is typically set to one or two.

2 Estimate the source s magnitude $|\mathbf{X}_s|$

 $|\hat{\mathbf{X}}_s| = \mathbf{M}_s \odot |\mathbf{X}|$

where \odot is an element-wise multiplication

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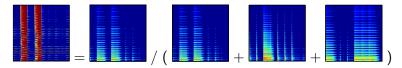
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Source Filtering II

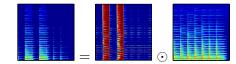
(

Example: Choose "D" pitches as a single source w/one basis vector

Compute filter
$$\mathbf{M}_s = \frac{\mathbf{w}_1 \mathbf{h}_1^{\mathrm{T}}}{\sum\limits_{i=1}^{K} \mathbf{w}_i \mathbf{h}_i^{\mathrm{T}}}$$
, with $\alpha = 1$



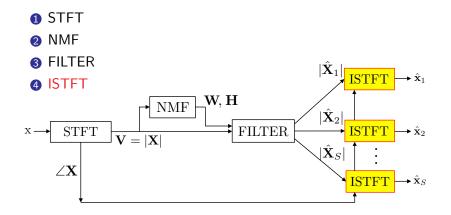
2 Multiply with $|\hat{\mathbf{X}}_s| = \mathbf{M}_s \odot |\mathbf{X}|$



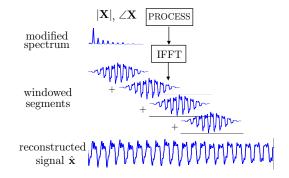
Source Filtering III

- \bullet The filter ${\bf M}$ is referred to as a masking filter or soft mask
- Tends to perform better than the reconstruction method
- Similar to Wiener filtering discussed in Talk 1

General Separation Pipeline

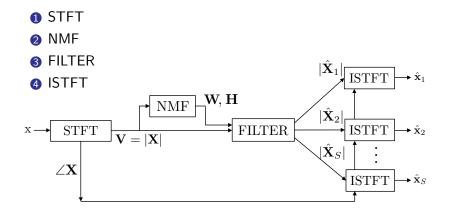


Inverse Short-Time Fourier Transform (ISTFT)



- Inputs $|\mathbf{X}|$ and phase $\angle \mathbf{X}$ matrices
- Outputs time domain signal **x**

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- Question: How do we solve for ${\bf W}$ and ${\bf H},$ given a known ${\bf V}?$
- Answer: Frame as optimization problem

 $\underset{\mathbf{W},\mathbf{H}\geq 0}{\text{minimize}} \ D(\mathbf{V} \,||\, \mathbf{W}\, \mathbf{H})$

where D is a measure of "divergence".

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Some choices for D:

- Euclidean: $D(\mathbf{V} || \hat{\mathbf{V}}) = \sum_{i,j} (\mathbf{V}_{ij} \hat{\mathbf{V}}_{ij})^2$
- Kullback-Leibler: $D(\mathbf{V} || \hat{\mathbf{V}}) = \sum_{i,j} \left(\mathbf{V}_{ij} \log \frac{\mathbf{V}_{ij}}{\hat{\mathbf{V}}_{ij}} - \mathbf{V}_{ij} + \hat{\mathbf{V}}_{ij} \right)$

We will focus on KL divergence in today's lecture.

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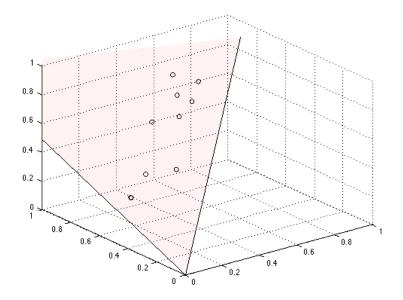
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Geometric View of NMF



How does one solve

$$\underset{\mathbf{W},\mathbf{H}\geq 0}{\text{minimize}} \quad \sum_{i,j} \left(\mathbf{V}_{ij} \log \frac{\mathbf{V}_{ij}}{(\mathbf{W} \mathbf{H})_{ij}} - \mathbf{V}_{ij} + (\mathbf{W} \mathbf{H})_{ij} \right)?$$

Tricks of the trade for minimizing a function $f(\mathbf{x})$.

- closed-form solutions: solve $\nabla f(\mathbf{x}) = 0$.
- gradient descent: iteratively move in steepest descent dir.

$$\mathbf{x}^{(\ell+1)} \leftarrow \mathbf{x}^{(\ell)} - \eta \nabla f(\mathbf{x}^{(\ell)}).$$

Newton's method: iteratively minimize quadratic approx.

$$\begin{split} \mathbf{x}^{(\ell+1)} &\leftarrow \underset{\mathbf{x}}{\operatorname{argmin}} f(\mathbf{x}^{(\ell)}) + \nabla f(\mathbf{x}^{(\ell)})^T (\mathbf{x} - \mathbf{x}^{(\ell)}) \\ &+ \frac{1}{2} (\mathbf{x} - \mathbf{x}^{(\ell)})^T \nabla^2 f(\mathbf{x}^{(\ell)}) (\mathbf{x} - \mathbf{x}^{(\ell)}) \end{split}$$

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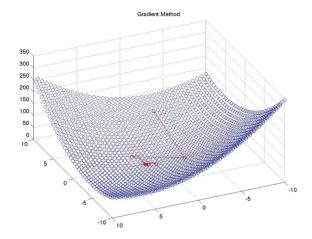
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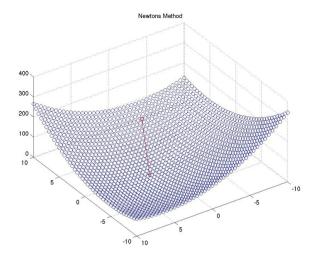
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Gradient Descent



Newton's Method



Coordinate descent

- Instead of minimizing $f(\mathbf{x})$, minimize $f(\mathbf{x}_i; \mathbf{x}_{-i}^{(\ell)})$ and cycle over i.
- Useful when $f(\mathbf{x}_i; \mathbf{x}_{-i}^{(\ell)})$ can be minimized in closed form.

Majorization-minimization

1 Find a majorizing function g for f at current iterate $\mathbf{x}^{(\ell)}$.

•
$$f(\mathbf{x}) < g(\mathbf{x}; \mathbf{x}^{(\ell)})$$
 for all $\mathbf{x} \neq \mathbf{x}^{(\ell)}$

•
$$f(\mathbf{x}^{(\ell)}) = g(\mathbf{x}^{(\ell)}; \mathbf{x}^{(\ell)})$$

Coordinate descent

- Instead of minimizing $f(\mathbf{x})$, minimize $f(\mathbf{x}_i; \mathbf{x}_{-i}^{(\ell)})$ and cycle over i.
- Useful when $f(\mathbf{x}_i; \mathbf{x}_{-i}^{(\ell)})$ can be minimized in closed form.

Majorization-minimization

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- 2 Minimize the majorizing function to obtain $\mathbf{x}^{(\ell+1)}$.

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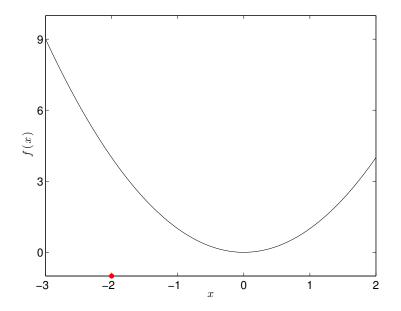
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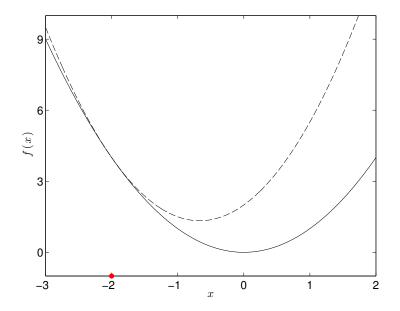
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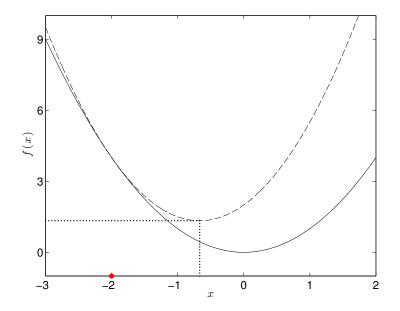
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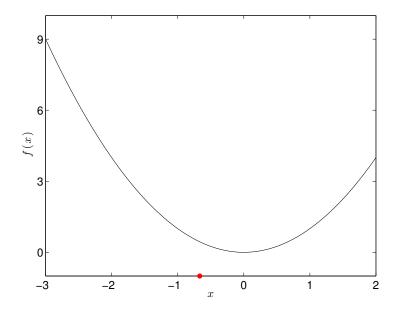
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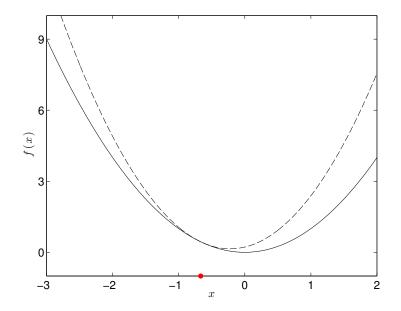




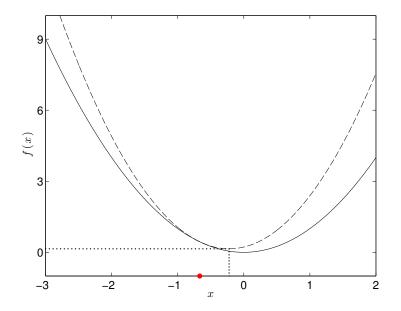




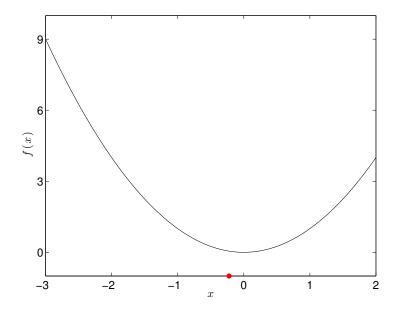
Majorization-minimization



Majorization-minimization



Majorization-minimization



To minimize

$$D(\mathbf{V} || \mathbf{W} \mathbf{H}) = \sum_{i,j} \left(\mathbf{V}_{ij} \log \frac{\mathbf{V}_{ij}}{(\mathbf{W} \mathbf{H})_{ij}} - \mathbf{V}_{ij} + (\mathbf{W} \mathbf{H})_{ij} \right)$$

$$\stackrel{\text{cst.}}{=} \sum_{i,j} - \mathbf{V}_{ij} \log \sum_{k} \mathbf{W}_{ik} \mathbf{H}_{kj} + \sum_{i,j} \sum_{k} \mathbf{W}_{ik} \mathbf{H}_{kj}$$

we use **(block) coordinate descent**: optimize \mathbf{H} for \mathbf{W} fixed, then optimize \mathbf{W} for \mathbf{H} fixed (rinse and repeat).

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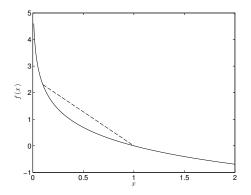
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Not quite, so let's try to majorize the function. A useful tool is **Jensen's inequality**, which says that for **convex** functions f:

 $f(average) \leq average \text{ of } f$



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To apply Jensen's inequality, we introduce weights $\sum_k \pi_{ijk} = 1.$

$$= \sum_{i,j} \left(-\mathbf{V}_{ij} \log \sum_{k} \pi_{ijk} \frac{\mathbf{W}_{ik} \mathbf{H}_{kj}}{\pi_{ijk}} + \sum_{k} \mathbf{W}_{ik} \mathbf{H}_{kj} \right)$$
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Now this function can be minimized exactly!

$$\mathbf{H}_{kj}^* = \frac{\sum_i \mathbf{V}_{ij} \, \pi_{ijk}}{\sum_i \mathbf{W}_{ik}}$$

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But I haven't told you what π_{ijk} is. We have to choose π_{ijk} to make the function a *majorizing* function.

$$\pi_{ijk} = \frac{\mathbf{W}_{ik} \mathbf{H}_{kj}^{(\ell)}}{\sum_{k} \mathbf{W}_{ik} \mathbf{H}_{kj}^{(\ell)}} \text{ does the trick.}$$

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If we substitute $\pi_{ijk} = \frac{\mathbf{W}_{ik} \mathbf{H}_{kj}^{(\ell)}}{\sum_{k} \mathbf{W}_{ik} \mathbf{H}_{kj}^{(\ell)}}$, we obtain the updates: $\mathbf{H}_{kj}^{(\ell+1)} \leftarrow \frac{\sum_{i} \mathbf{V}_{ij} \frac{\mathbf{W}_{ik} \mathbf{H}_{kj}^{(\ell)}}{\sum_{k} \mathbf{W}_{ik} \mathbf{H}_{kj}^{(\ell)}}}{\sum_{i} \mathbf{W}_{ik}}$ $= \mathbf{H}_{kj}^{(\ell)} \cdot \frac{\sum_{i} \left(\frac{\mathbf{V}}{\mathbf{W} \mathbf{H}^{(\ell)}}\right)_{ij} \mathbf{W}_{ik}}{\sum_{i} \mathbf{W}_{ik}}$

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Using $D(\mathbf{V} || \mathbf{W} \mathbf{H}) = D(\mathbf{V}^T || \mathbf{H}^T \mathbf{W}^T)$, we obtain a similar update for \mathbf{W} .

Now we just iterate between:

• Updating W.

- **2** Updating **H**.
- **3** Checking $D(\mathbf{V} || \mathbf{W} \mathbf{H})$. If the change since the last iteration is small, then declare convergence.

The algorithm is summarized below:

Algorithm KL-NMF

 $\begin{array}{c} \text{initialize } \mathbf{W}, \mathbf{H} \\ \text{repeat} \\ \mathbf{H} \leftarrow \mathbf{H} . ^* \frac{\mathbf{W}^T \frac{\mathbf{V}}{\mathbf{W}\mathbf{H}}}{\mathbf{W}^T \mathbf{1}} \\ \mathbf{W} \leftarrow \mathbf{W} . ^* \frac{\mathbf{V}}{\mathbf{W}\mathbf{H}} \frac{\mathbf{H}^T}{\mathbf{1}\mathbf{H}^T} \\ \end{array}$

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initialize W, H repeat $\mathbf{H} \leftarrow \mathbf{H} \cdot^* \frac{\mathbf{W}^T \cdot \mathbf{V}}{\mathbf{W}^T \mathbf{1}}$ $\mathbf{W} \leftarrow \mathbf{W} \cdot^* \frac{\mathbf{V}}{\mathbf{W}\mathbf{H}} \frac{\mathbf{H}^T}{\mathbf{1}\mathbf{H}^T}$ until convergence return W

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until convergence return W, H

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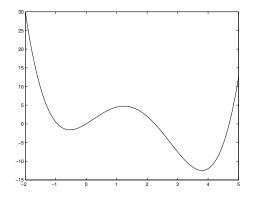
- 1 Updating W.
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The algorithm is summarized below:

 $\label{eq:constraint} \begin{array}{l} \hline \textbf{Algorithm} \quad \text{KL-NMF} \\ \hline \textbf{initialize} \ \textbf{W}, \textbf{H} \\ \textbf{repeat} \\ \textbf{H} \leftarrow \textbf{H} .^* \frac{\textbf{W}^T \frac{\textbf{V}}{\textbf{W}\textbf{H}}}{\textbf{W}^T \textbf{1}} \\ \textbf{W} \leftarrow \textbf{W} .^* \frac{\frac{\textbf{V}}{\textbf{W}\textbf{H}} \textbf{H}^T}{\textbf{1}\textbf{H}^T} \\ \textbf{until convergence} \ \textbf{return} \ \textbf{W}, \textbf{H} \end{array}$

Caveats

• The NMF problem is **non-convex**.



- The algorithm is only guaranteed to find a local optimum.
- The algorithm is sensitive to choice of initialization.

Roadmap of Talk

1 Motivation

- **2** Current Approaches
- 3 Non-Negative Matrix Factorization (NMF)
- **4** Source Separation via NMF
- **6** Algorithms for NMF
- 6 Matlab Code

STFT

```
FFTSIZE = 1024;
HOPSIZE = 256;
WINDOWSIZE = 512;
X = myspectrogram(x,FFTSIZE,fs,hann(WINDOWSIZE),-HOPSIZE);
V = abs(X(1:(FFTSIZE/2+1),:));
F = size(V,1);
T = size(V,2);
```

- https://ccrma.stanford.edu/~jos/sasp/Matlab_ listing_myspectrogram_m.html
- https://ccrma.stanford.edu/~jos/sasp/Matlab_ listing_invmyspectrogram_m.html

NMF

```
function [W, H] = nmf(V, K, MAXITER)
F = size(V, 1):
T = size(V, 2);
rand('seed'.0)
W = 1 + rand(F, K);
H = 1 + rand(K, T);
ONES = ones(F,T);
for i=1:MAXITER
    % update activations
    H = H .* (W'*( V./(W*H+eps))) ./ (W'*ONES);
    % update dictionaries
    W = W .* ((V./(W*H+eps))*H') ./(ONES*H');
end
% normalize W to sum to 1
sumW = sum(W);
W = W * diag(1./sumW);
H = diag(sumW) * H;
```

FILTER & ISTFT

```
phi = angle(X);
% reconstruct each basis as a separate source
for i=1:K
```

```
XmagHat = W(:,i)*H(i,:);
```

% create upper half of frequency before istft
XmagHat = [XmagHat; conj(XmagHat(end-1:-1:2,:))];

```
% Multiply with phase
XHat = XmagHat.*exp(1i*phi);
```

```
xhat(:,i) = real(invmyspectrogram(XHat,HOPSIZE))';
```

end

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