

MUS420 Supplement  
 Stability Proof for a Cylindrical Bore with Conical Cap

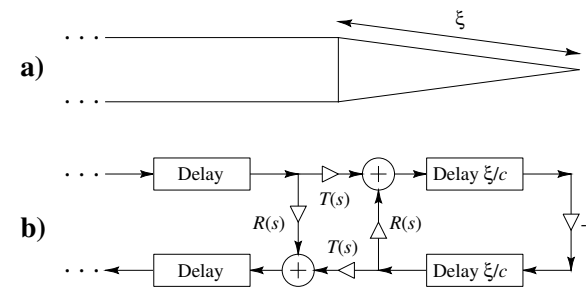
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**Outline**

- Cylinder with Conical Cap
- Scattering Filters at the Cylinder-Cone Junction
- Reflectance of the Conical Cap
- Reflectance of Conical Cap Seen from Cylinder
- Stability Proof
- Poles at DC

**Cylinder with Conical Cap**



- Cylinder open or closed on left side
- Otherwise closed
- Obviously passive physically
- Hard to show! [ $R(s)$  and  $T(s)$  are unstable]

**Scattering Filters at the Cylinder-Cone Junction**

Wave impedance at frequency  $\omega$  rad/sec in a converging cone:

$$Z_{\xi}(j\omega) = \frac{\rho c}{S(\xi)} \cdot \frac{j\omega}{j\omega - c/\xi} \quad (\text{converging cone impedance})$$

where

$$\begin{aligned}\xi &= \text{distance to the apex of the cone} \\ S(\xi) &= \text{cross-sectional area of cone} \\ \rho c &= \text{wave impedance in open air}\end{aligned}$$

In the limit as  $\xi \rightarrow \infty$ ,

$$Z_\infty(j\omega) = \frac{\rho c}{S} \quad (\text{cylindrical tube impedance})$$

Reflectance of the conical cap, seen from cylinder:

$$R(s) = -\frac{c/\xi}{c/\xi - 2s}$$

Transmittance to the right:

$$T(s) = 1 + R(s) = -\frac{2s}{c/\xi - 2s}$$

- $R(s)$  and  $T(s)$  are first-order transfer functions, each having a single real pole at  $s = c/(2\xi) \Rightarrow$  *unstable*
- $R(s)$  and  $T(s)$  identical from left and right given no wavefront area discontinuity.

## Reflectance of the Conical Cap

- Let  $t_\xi \triangleq \xi/c$  denote the time to propagate across the length of the cone in one direction
- Reflectance of complete (lossless) cone is  $-1$  for pressure waves (reflects like an open-ended cylinder)
- Round-trip transfer function from cone entrance to tip and back is

$$R_{t_\xi}(s) \triangleq -e^{-2st_\xi} = e^{-st_\xi}(-1)e^{-st_\xi}$$

(reflectance seen *inside* the cone)

## Reflectance of Conical Cap Seen from Cylinder

From the figure, we can derive the conical cap reflectance

to be

$$\begin{aligned}
 R_J(s) &= \frac{R(s) + 2R(s)R_{t_\xi}(s) + R_{t_\xi}(s)}{1 - R(s)R_{t_\xi}(s)} \\
 &= \frac{1 + (1 + 2st_\xi)R_{t_\xi}(s)}{2st_\xi - 1 - R_{t_\xi}(s)} \\
 &= \frac{1 - (1 + 2st_\xi)e^{-2st_\xi}}{2st_\xi - 1 + e^{-2st_\xi}} \\
 &\triangleq \frac{N(s)}{D(s)}
 \end{aligned}$$

For very large  $t_\xi$ , the conical cap reflectance approaches  $R_J = -e^{-2st_\xi}$  which coincides with the impedance of a length  $\xi = ct_\xi$  open-end cylinder, as expected.

### Stability Proof Outline

- A transfer function  $R_J(s) = N(s)/D(s)$  is stable if there are no poles in the right-half  $s$  plane. That is, for each zero  $s_i$  of  $D(s)$ , we must have  $\text{re}\{s_i\} \leq 0$ . If this can be shown, along with  $|R_J(j\omega)| \leq 1$ , then the reflectance  $R_J$  is shown to be passive.
- We must also study the system zeros (roots of  $N(s)$ ) in order to determine if there are any pole-zero cancellations (common factors in  $D(s)$  and  $N(s)$ ).

- Since  $\text{re}\{st_\xi\} \geq 0$  if and only if  $\text{re}\{s\} \geq 0$ , for  $t_\xi > 0$ , we may set  $t_\xi = 1$  without loss of generality. Thus, we need only study the roots of

$$\begin{aligned}
 N(s) &= 1 - e^{-2s} - 2se^{-2s} \\
 D(s) &= 2s - 1 + e^{-2s}
 \end{aligned}$$

If this system is stable, we have stability also for all  $t_\xi > 0$ .

- Since  $e^{-2s}$  is not a rational function of  $s$ , the reflectance  $R_J(s)$  may have infinitely many poles and zeros.

### Stability Proof

First consider the roots of the denominator

$$D(s) = 2s - 1 + e^{-2s}.$$

At any pole (solution  $s$  of  $D(s) = 0$ ), we must have

$$s = \frac{1 - e^{-2s}}{2}$$

To obtain separate equations for the real and imaginary parts, take the real and imaginary parts of  $D(\sigma + j\omega) = 0$  to get

$$\begin{aligned}
 \text{re}\{D(s)\} &= (2\sigma - 1) + e^{-2\sigma} \cos(2\omega) = 0 \\
 \text{im}\{D(s)\} &= 2\omega - e^{-2\sigma} \sin(2\omega) = 0
 \end{aligned}$$

Both of these equations must hold at any pole of the reflectance. For stability, we further require  $\sigma \leq 0$ . Defining  $\tau = 2\sigma$  and  $\nu = 2\omega$ , we obtain the simpler conditions

$$\begin{aligned} e^\tau(1 - \tau) &= \cos(\nu) \\ e^\tau &= \frac{\sin(\nu)}{\nu} \end{aligned}$$

For any poles of  $R_{vJ}(s)$  on the  $j\omega$  axis, we have  $\tau = 0$ , and the second equation reduces to  $\text{sinc}(\nu) = 1$ . It is well known that the sinc function is less than 1 in magnitude at all  $\nu$  except  $\nu = 0$ . Therefore, this relation can hold only at  $\omega = \nu = 0$ , and so

*Any right-half-plane poles occur at  $\omega = 0$ .*

### Stability Proof, continued

The same argument can be extended to the entire right-half plane as follows. Going back to

$$\frac{\sin(\nu)}{\nu} = e^\tau,$$

since  $|\sin(\nu)/\nu| \leq 1$  for all real  $\nu$ , and since  $|e^\tau| > 1$  for  $\tau > 0$ , this equation clearly has no solutions in the right-half plane. Therefore,

*Any right-half-plane poles occur at  $s = 0$ .*

### A Pole at DC

Since both of the conditions

$$\begin{aligned} e^\tau(1 - \tau) &= \cos(\nu) \\ e^\tau &= \frac{\sin(\nu)}{\nu} \end{aligned}$$

are clearly satisfied for  $\tau = \nu = 0$ , we see that there is in fact a pole in the reflectance at dc ( $s = 0$ ), provided it is not canceled by a zero at dc in the numerator  $N(s)$ .

### The Left-Half Plane

In the left-half plane, there are many potential poles:

- The first of the two equations

$$e^\tau(1 - \tau) = \cos(\nu)$$

has infinitely many solutions for each  $\tau < 0$ , since the elementary inequality  $1 - \tau \leq e^{-\tau}$  implies

$$e^\tau(1 - \tau) < e^\tau e^{-\tau} = 1$$

- The second equation,

$$e^\tau = \frac{\sin(\nu)}{\nu}$$

has an increasing number of solutions as  $\tau$  grows more and more negative.

- As  $\tau \rightarrow -\infty$ , the number of solutions becomes infinite and are given by the zeros of  $\sin(\nu)$
- At  $\tau \rightarrow -\infty$ , the solutions of the other equation converge to the zeros of  $\cos(\nu)$
- Thus, the solutions of

$$e^\tau(1 - \tau) = \cos(\nu)$$

$$e^\tau = \frac{\sin(\nu)}{\nu}$$

may not necessarily occur together for  $\tau < 0$ , as they must.

### Poles at $s=0$

We know from the foregoing that the denominator of the cone reflectance has at least one root at  $s = 0$ . We now investigate the “dc behavior” more thoroughly.

- A hasty analysis based on the reflection and transmission filters (see figure) might conclude that the reflectance of the conical cap converges to  $-1$  at dc, since  $R(0) = -1$  and  $T(0) = 0$ . However, this is incorrect.

- Instead, it is necessary to take the limit as  $\omega \rightarrow 0$  of the complete conical cap reflectance  $R_J(s)$ :

$$R_J(s) = \frac{1 - e^{-2s} - 2se^{-2s}}{2s - 1 + e^{-2s}}$$

We already discovered a root at  $s = 0$  in the denominator in the context of the preceding stability proof. However, note that the numerator also goes to zero at  $s = 0$ . This indicates a pole-zero cancellation at dc.

- To find the reflectance at dc, we may use L'Hospital's rule to obtain

$$R_J(0) = \lim_{s \rightarrow 0} \frac{N'(s)}{D'(s)} = \lim_{s \rightarrow 0} \frac{4se^{-2s}}{2 - 2e^{-2s}}$$

and once again the limit is an indeterminate  $0/0$  form.

- We apply L'Hospital's rule again to obtain

$$R_J(0) = \lim_{s \rightarrow 0} \frac{N''(s)}{D''(s)} = \lim_{s \rightarrow 0} \frac{(4 - 8s)e^{-2s}}{4e^{-2s}} = +1$$

Thus, two poles and zeros cancel at dc, and the dc reflectance is  $+1$ , not  $-1$  as an analysis based only on the scattering filters would indicate.

- From a physical point of view, it makes more sense that the cone should “look like” a simple rigid

termination of the cylinder at dc, since its length becomes vanishingly small compared with the wavelength in the limit.

- Another method of showing this result is to form a Taylor series expansion of the numerator and denominator:

$$N(s) = 2s^2 - \frac{8s^3}{3} + 2s^4 + \dots$$
$$D(s) = 2s^2 - \frac{4s^3}{3} + \frac{2s^4}{3} + \dots$$

Both series begin with the term  $2s^2$  which means both the numerator and denominator have two roots at  $s = 0$ . Hence, again the conclusion is two pole-zero cancellations at dc.

- The series for the conical cap reflectance is

$$R_J(s) = 1 - \frac{2s}{3} + \frac{2s^2}{9} - \frac{4s^3}{135} + \frac{2s^4}{405} + \dots$$

which approaches +1 as  $s \rightarrow 0$ .