Abstract

A new model for transverse piano string vibration, second order in time, which models frequency-dependent loss and dispersion effects is presented here. It is then shown how a digital waveguide structure may be related directly to this model. Finally, the model parameters are fit to experimental data from a grand piano.

1 Introduction

Several models of transverse wave propagation on a piano string have appeared in the literature (Chaigne and Askenfelt 1994a; Chaigne and Askenfelt 1994b; Chaigne 1992). These models are always framed in terms of a partial differential equation (PDE); usually, the starting point for such a model is the wave equation (Fletcher and Rossing 1991), and more realistic features, such as dispersion and frequency-dependent loss are incorporated through several perturbation terms. The most advanced such model (Chaigne and Askenfelt 1994a) has been used as the basis for a high-fidelity sound synthesis technique (Chaigne and Askenfelt 1994b).

Digital waveguides (Smith 1987; Karjalainen, Välimäki, and Tolonen 1998) are filter-like structures which model wave propagation as purely lossless throughout the length of the string, with loss and dispersion lumped in terminating filters. They are thus simulations of modified physical systems, but are very efficient in the context of musical sound synthesis. The aim of this paper is to link PDE models and digital waveguides, and to explicitly show the relationship between the lumped filters used to model loss and dispersion and the parameters which define our PDE, which is a variant of Chaigne and Askenfelt’s system. Calibration of the filters to measured data is also discussed.

2 PDE model of a stiff string, with frequency-dependent loss

In this section, we present a new model of piano string vibration, which can be written as

$$\frac{\partial^2 y}{\partial t^2} = c^2 \frac{\partial^2 y}{\partial x^2} - \kappa^2 \frac{\partial^4 y}{\partial x^4} - 2b_1 \frac{\partial y}{\partial t} + 2b_2 \frac{\partial^3 y}{\partial x^2 \partial t}$$

(1)

Here, $y(x,t)$ is the transverse displacement of the string in a single plane, as a function of time $t \geq 0$ and position $x \in [0, L]$ where $L$ is the string length. The first term on the right-hand side of the equation, in the absence of the others, gives rise to wave like motion, with speed $c$. The second “ideal bar” term introduces dispersion, or frequency-dependent wave velocity, and is parametrized by a stiffness coefficient $\kappa$. The third and fourth terms allow for loss, and if $b_2 \neq 0$, decay rates will be frequency-dependent. A complete model is obtained by including a hammer excitation term $f(x,t)$, possibly including nonlinear effects, on the right-hand side, and supplying a realistic set of initial and boundary conditions.

If the term $2b_2 \frac{\partial^3 y}{\partial x^2 \partial t}$ is replaced by $2b_2 \frac{\partial^3 y}{\partial x^2 \partial t}$, we then arrive at the Chaigne and Askenfelt model, and thus the distinction is solely in the modeling of frequency-dependent loss. In a subsequent paper, we will describe in detail the reasons why the model presented here is preferable, but we briefly summarize them here: First, the restriction to second order of the time derivatives greatly simplifies analysis and allows the construction of more efficient finite difference schemes for which numerical stability is easily verifiable. Second, because the number of independent wave-like solutions is reduced to two, it becomes possible to identify this equation directly with digital waveguide models which are based on the use of bidirectional delay lines which propagate traveling waves in opposite directions; for the Chaigne model, third order in time, there are three independent solutions. Finally, for a second-order equation such as (1), generalization to more accurate models of dispersion and loss (through the addition of more terms to the PDE) can be achieved without disrupting the well-posedness of the system (Strikwerda 1989), provided certain very simple conditions are respected. We note that in a later publication (Chaigne and Doutaut...
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solutions of the form
linear and shift-invariant with respect to both time and
be reintroduced subsequently. Because the PDE (1) is
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space, it can be analyzed by considering wave-like so-
ical processes (in particular loss), and which may be
usefully be applied to the analysis and synthesis of mu-
sical tones.

3 An associated waveguide model

3.1 Dispersion and loss

For the purpose of examining the dispersion and loss characteristics of the traveling-wave solutions, the assumption of a string of infinite length is permissible and simplifies the analysis; boundary conditions may be reintroduced subsequently. Because the PDE (1) is linear and shift-invariant with respect to both time and space, it can be analyzed by considering wave-like solutions of the form

\[ y(x,t) = e^{at + jsx} \]  (2)

Here, \( \beta \) is the real spatial wavenumber, and \( s \) is a complex frequency variable. If such a solution is inserted in (1), then a dispersion relation (Elmore and Heald 1969) results:

\[ s^2 + 2(b_1 + b_2 \beta^2)s + c^2 \beta^2 + \kappa^2 \beta^2 = 0 \]  (3)

(The analysis of the propagation characteristics of a single wave-like solution can be considered to be a shortcut to the full Fourier and Laplace analysis, which yields the same resulting equation.) The roots of this equation, \( s_+ \) and \( s_- \), are thus

\[ s_{\pm} = -b_1 - b_2 \beta^2 \pm \sqrt{(b_1 + b_2 \beta^2)^2 - c^2 \beta^2 - \kappa^2 \beta^4} \]  (4)

Over the range of \( \beta \) for which the quantity under the radical is negative (for \( b_1 \) and \( b_2 \) small, this will be true for a substantial range of wavenumbers), \( s_{\pm} \) form a complex conjugate pair, and it is then possible to separate the real and imaginary parts in terms of \( \beta \) by writing \( s_{\pm} = \eta \pm j\omega \), with

\[ \eta = -b_1 - b_2 \beta^2 \]  (5a)
\[ \omega = \sqrt{c^2 \beta^2 + \kappa^2 \beta^4 - (b_1 + b_2 \beta^2)^2} \]  (5b)

\( \eta \) is to be interpreted as a frequency-dependent loss parameter (notice that for \( b_1 \) and \( b_2 \) chosen non-negative, we have \( \eta \leq 0 \) for all \( \beta \), so exponential solutions are always damped), and \( \omega \) is a real frequency variable.

3.2 The corresponding waveguide filter

We now turn to the problem of relating digital waveguide parameters (to be discussed shortly) to the loss and inharmonicity parameters discussed in Section 3.1. To this end, we now return to the string of length \( L \), under fixed boundary conditions. A simplified digital waveguide structure which can be used for piano synthesis is as shown in Figure 1.

![Waveguide filter structure.](image)

The excitation \( I \) corresponds to the energy supplied to the string initially at one endpoint (by the hammer). The output \( O \) will then represent the time waveform of the string’s motion at the bridge, or perhaps a unidirectional traveling-wave component at a selected point along the string. We will denote by \( D \) the time taken for energy in the lowest mode to complete a round-trip passage of distance \( 2L \) over the string; this minimum delay is simulated by a digital delay line of duration \( D_0 \) (as shown in Figure 1). The lumped digital filter \( F \) simulates the remaining accumulated lossy and dispersive effects over the same round-trip propagation distance; in particular it accounts for losses in the string itself (from air friction, and viscoelastic effects), as well as losses at the endpoints (where energy is transmitted into the soundboard).

Consider an exponential wave solution propagating along the string, over a time duration \( D \). From the definition of the exponential solution from (2), the propagation can be represented by a multiplicative phase factor \( \exp(sD + 2j\beta L) \). The modulus and phase of this factor are thus related to the filter \( F \) by

\[ |F| = e^{\eta D} = e^{-b_1 D - b_2 D \beta^2} \]  (6)
\[ \arg(F) = \omega D - 2\beta L \]  (7)

Thus the phase of \( F \) is that of the multiplicative phase factor, except for the removal of the constant delay

In order to rewrite this filter in terms of the frequency \( \omega \), it is necessary to express the wavenumber \( \beta \) in terms of \( \omega \). From (5b), and solving for \( \beta \), the required expression will be

\[ \beta = \pm \sqrt{-q \pm \sqrt{q^2 + 4r(b_1^2 + \omega^2)}} \]  (8)
with
\[ q = c^2 - 2 b_1 b_2 \quad r = \kappa^2 - b_2^2 \quad (9) \]

Given that, for realistic piano string modeling, \( b_1 \approx 1 \) and \( b_2 \approx 10^{-4} \), we make the simplifying assumptions
\[ b_1 b_2 \ll c^2 \quad b_2^2 \ll \kappa^2 \quad b_1^2 \ll \omega^2 \quad (10) \]
which permit the following approximation of \( \beta \):
\[ \beta \approx \pm \sqrt{\frac{c^2 \gamma}{2 \kappa}} \quad (11) \]
where
\[ \gamma = \sqrt{1 + 4 \kappa^2 \omega^2 / c^4} - 1 \quad (12) \]

Finally, we arrive at approximate expressions for the modulus and phase of the filter \( F \) as a function of frequency \( \omega \),
\[ |F| \approx \exp \left( -D \left[ b_1 + \frac{b_2 c^2 \gamma}{2 \kappa^2} \right] \right) \quad (13a) \]
\[ \arg(F) \approx \omega D - \frac{Lc \sqrt{2} \gamma}{\kappa} \quad (13b) \]

4 Calibration of loss and inharmonicity parameters from experimental data

Because this model is intended for use in the context of musical sound synthesis, we here discuss the calibration of \( b_1 \) and \( b_2 \), and the determination of the stiffness parameter. To this end, many measurements were taken, using a Yamaha Disklavier; for each note, the vibration of the string at the bridge was measured (using an accelerometer), for a hammer speed of 2.1 m/s.

Using signal processing techniques (Aramaki, Bensa, Daudet, Guillemain, and Kronland-Martinet 2002), the damping coefficients were then determined. Over most of the piano range, however, the hammer strikes not one, but two or three strings simultaneously. The coupling gives rise to perceptually significant phenomena such as beating and two-stage decay; these effects are not taken into account in model (1). For these multi-string notes, the calculated damping coefficients can be thought of describing the global perceived decay of the sound. For each note, \( b_1 \) and \( b_2 \) were calculated from (13a). The evolution of these parameters as a function of MIDI note number is shown in Figure 2.

\[ b_1 = 7.4 \times 10^{-3} f_0 - 4.49 \times 10^{-2} \quad (14a) \]
\[ b_2 = 1.0 \times 10^{-6} f_0 + 1.07 \times 10^{-5} \quad (14b) \]

These simple empirical descriptions of \( b_1 \) and \( b_2 \) allow the reproduction of piano tones whose damping will be identical to that of the perceived acoustic note. The study of more detailed models of multi-string coupling is currently in progress; a multi-string waveguide model was presented in (Aramaki, Bensa, Daudet, Guillemain, and Kronland-Martinet 2002).

In practice, it is helpful to work with more perceptually significant parameters. We may express the phase of the filter \( F \) in terms of the inharmonicity coefficient \( B \) (H. Fletcher and Stratton 1962) and the fundamental frequency \( \omega_0 \) by
\[ B = \kappa^2 \omega_0^2 / c^4 \quad c = \omega_0 L / \pi \quad (15) \]
from which, the phase of \( F \) may be rewritten as
\[ \arg(F) \approx \omega D - \frac{\pi \sqrt{B}}{\sqrt{B}} \sqrt{-1 + \sqrt{1 + 4 B \omega^2 / \omega_0^2}} \quad (16) \]

This expression for the phase allows the estimation of \( \omega_0 \) and \( B \) for each note. The inharmonicity factor \( B \)
is plotted as a function of MIDI note number in Figure 3; $B$ is an increasing function of the note number, except over the bass range, where the strings are double-wrapped.

The determination of $b_1$, $b_2$, $B$ and $\omega_0$ for each note allows us, then, an explicit expression of the behavior of the filter $F$ as a function of note number. In order to represent the evolution of the loop filter in terms of notes, we show in Figure 4 the modulus and phase of the elementary filter $\delta F$, normalized with respect to the filter $D$:

$$\delta F = F_1/D$$

The modulus, which also takes into account the losses at the endpoints, is decreasing with the note number. But we can notice that this behavior is slightly different for the wrapped strings (A0, A1) than for the other strings (A2, E3, A3). It is worth noting that although $B$ is an increasing function of note number, the phase of the filters $\delta F$ grows less rapidly. This is due to the fact that the phase of the filter depends not only on $B$, but also the fundamental frequency. This effect can be understood by expanding the expression for the phase of $F$ for $4B(\omega/\omega_0)$ near zero; one gets

$$\arg(F) \approx \frac{\omega^2 B}{2\omega_0^2}$$

Though meaningful only for the first few partials, one can clearly see the dependence which is decreasing with the fundamental frequency.

5 Conclusions

We have presented in this paper a new model of piano string vibration, which takes into account effects of stiffness and frequency-dependent loss. This model can be considered to be a variant of the model of Chaigne and Askenfelt which possesses the advantage that explicit solutions may be derived and related to a digital waveguide structure, to be used for musical sound synthesis purposes. We also have shown how the parameters which define the model may be calibrated to experimental data and have provided a simple description of the variation of these parameters over the musical range of the piano.

Future work will involve extending the identification with a PDE to the multi-string framework.

References


ter for Computer Research in Music and Acoustics (CCRMA), Department of Music, Stanford University. See also http://www-ccrma.stanford.edu/~jos/wg.html.